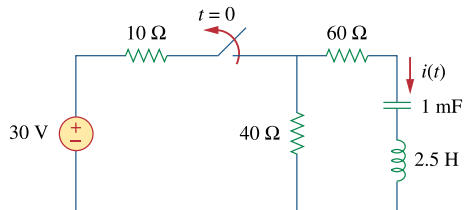


EECS70A / CSE 70A Network Analysis I  
Prof. Peter Burke

Homework # 6 solution

Q1. Problem 8.16



$$\text{At } t = 0, \quad i(0) = 0, \quad V_C(0) = 40 \times 30 / 50 = 24\text{V}$$

For  $t > 0$ , we have a source-free RLC circuit.

$$\alpha = R / (2L) = (40 + 60) / 5 = 20 \quad \text{and} \quad \omega_o = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{10^{-3} \times 2.5}} = 20$$

$\omega_o = \alpha$  leads to critical damping

$$i(t) = [(A + Bt)e^{-20t}], \quad i(0) = 0 = A$$

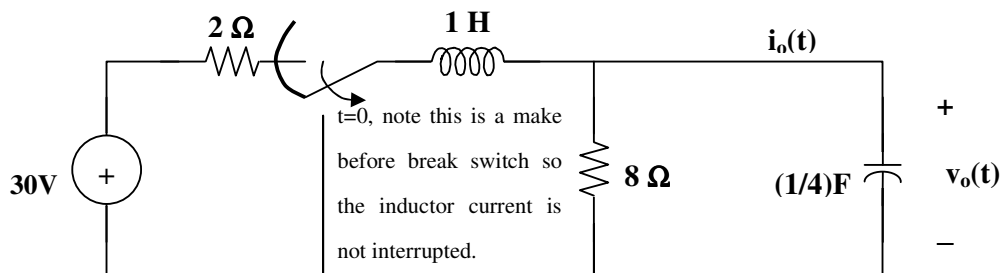
$$di/dt = \{[Be^{-20t}] + [-20(Bt)e^{-20t}]\},$$

$$\text{but } di(0)/dt = -(1/L)[Ri(0) + V_C(0)] = -(1/2.5)[0 + 24]$$

$$\text{Hence, } B = -9.6 \text{ or } i(t) = \underline{\underline{[-9.6te^{-20t}] \text{ A}}}$$

Q2. Problem 8.25:

In the circuit in Fig. 8.76, calculate  $i_o(t)$  and  $v_o(t)$  for  $t > 0$ .



At  $t = 0^-$ ,  $v_o(0) = (8/(2 + 8))(30) = 24$

For  $t > 0$ , we have a source-free parallel RLC circuit.

$$\alpha = 1/(2RC) = 1/4$$

$$\omega_o = 1/\sqrt{LC} = 1/\sqrt{1 \times 1/4} = 2$$

Since  $\alpha$  is less than  $\omega_o$ , we have an under-damped response.

$$\omega_d = \sqrt{\omega_o^2 - \alpha^2} = \sqrt{4 - (1/16)} = 1.9843$$

$$v_o(t) = (A_1 \cos \omega_d t + A_2 \sin \omega_d t) e^{-\alpha t}$$

$$v_o(0) = 30(8/(2+8)) = 24 = A_1 \text{ and } i_o(t) = C(dv_o/dt) = 0 \text{ when } t = 0.$$

$$dv_o/dt = -\alpha(A_1 \cos \omega_d t + A_2 \sin \omega_d t) e^{-\alpha t} + (-\omega_d A_1 \sin \omega_d t + \omega_d A_2 \cos \omega_d t) e^{-\alpha t}$$

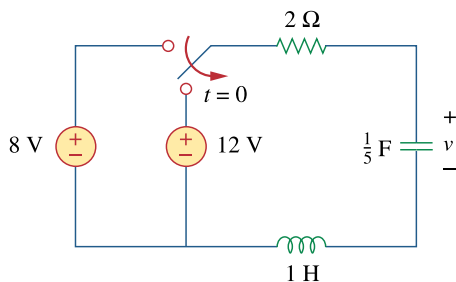
$$\text{at } t = 0, \text{ we get } dv_o(0)/dt = 0 = -\alpha A_1 + \omega_d A_2$$

$$\text{Thus, } A_2 = (\alpha/\omega_d)A_1 = (1/4)(24)/1.9843 = 3.024$$

$$v_o(t) = \underline{\underline{(24 \cos 1.9843t + 3.024 \sin 1.9843t) e^{-t/4} \text{ volts.}}}$$

$$\begin{aligned} i_o(t) &= Cdv/dt = 0.25[-24(1.9843)\sin 1.9843t + \\ &3.024(1.9843)\cos 1.9843t - 0.25(24 \cos 1.9843t) - \\ &0.25(3.024 \sin 1.9843t)]e^{-t/4} \\ &= \underline{\underline{[0.000131 \cos 1.9843t - 12.095 \sin 1.9843t]e^{-t/4} \text{ A.}}} \end{aligned}$$

Q3. Problem 8.35



At  $t = 0^-$ ,  $i_L(0) = 0$ ,  $v(0) = v_C(0) = 8 \text{ V}$

For  $t > 0$ , we have a series RLC circuit with a step input.

$$\alpha = R/(2L) = 2/2 = 1,$$

$$\omega_o = 1/\sqrt{LC} = 1/\sqrt{1/5} = \sqrt{5}$$

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_o^2} = -1 \pm j2$$

$$v(t) = V_s + [(A\cos 2t + B\sin 2t)e^{-t}], \quad V_s = 12.$$

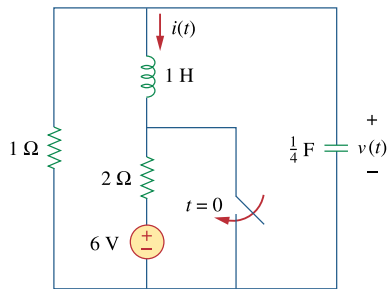
$$v(0) = 8 = 12 + A \quad \text{or } A = -4, \quad i(0) = Cdv(0)/dt = 0.$$

$$\text{But } dv/dt = [-(A\cos 2t + B\sin 2t)e^{-t}] + [2(-A\sin 2t + B\cos 2t)e^{-t}]$$

$$0 = dv(0)/dt = -A + 2B \quad \text{or } 2B = A = -4 \text{ and } B = -2$$

$$v(t) = \underline{\underline{\{12 - (4\cos 2t + 2\sin 2t)e^{-t}\} \text{ V.}}}$$

#### Q4. Problem 8.48



For  $t = 0^-$ , we obtain  $i(0) = -6/(1 + 2) = -2$  and  $v(0) = 2 \times 1 = 2$ .

For  $t > 0$ , the voltage is short-circuited and we have a source-free parallel RLC circuit.

$$\alpha = 1/(2RC) = (1)/(2 \times 1 \times 0.25) = 2$$

$$\omega_0 = 1/\sqrt{LC} = 1/\sqrt{1 \times 0.25} = 2$$

Since  $\alpha = \omega_0$ , we have a critically damped response.

$$s_{1,2} = -2$$

$$\text{Thus, } i(t) = [(A + Bt)e^{-2t}], \quad i(0) = -2 = A$$

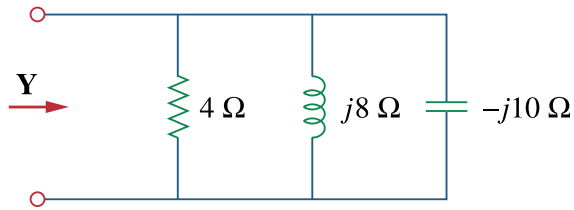
$$v = Ldi/dt = [Be^{-2t}] + [-2(-2 + Bt)e^{-2t}]$$

$$v_o(0) = 2 = B + 4 \quad \text{or } B = -2$$

$$\text{Thus, } i(t) = \underline{\underline{\{(-2 - 2t)e^{-2t}\} \text{ A}}}$$

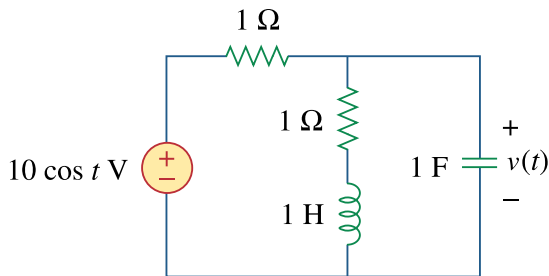
$$\text{And } v(t) = \underline{\underline{\{(2 + 4t)e^{-2t}\} \text{ V}}}$$

Q5. Problem 9.37



$$Y = \frac{1}{4} + \frac{1}{j8} + \frac{1}{-j10} = \underline{0.25 - j0.025 \text{ S}} = \underline{250 - j25 \text{ mS}}$$

Q6. Problem 9.41



$$\omega = 1,$$

$$1 \text{ H} \longrightarrow j\omega L = j(1)(1) = j$$

$$1 \text{ F} \longrightarrow \frac{1}{j\omega C} = \frac{1}{j(1)(1)} = -j$$

$$Z = 1 + (1 + j) \parallel (-j) = 1 + \frac{-j+1}{1} = 2 - j$$

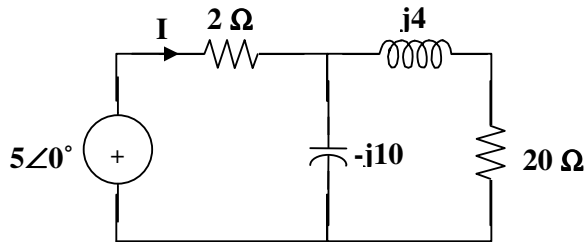
$$\mathbf{I} = \frac{\mathbf{V}_s}{\mathbf{Z}} = \frac{10}{2 - j}, \quad \mathbf{I}_c = (1 + j)\mathbf{I}$$

$$\mathbf{V} = (-j)(1 + j)\mathbf{I} = (1 - j)\mathbf{I} = \frac{(1 - j)(10)}{2 - j} = 6.325 \angle -18.43^\circ$$

$$\text{Thus, } v(t) = \underline{6.325 \cos(t - 18.43^\circ) \text{ V}}$$

Q6. Problem 9.47

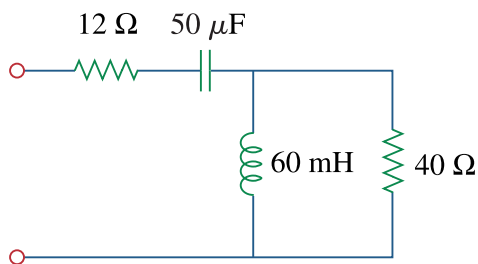
First, we convert the circuit into the frequency domain.



$$I_x = \frac{5}{2 + \frac{-j10(20 + j4)}{-j10 + 20 + j4}} = \frac{5}{2 + 4.588 - j8.626} = \frac{5}{10.854 \angle -52.63^\circ} = 0.4607 \angle 52.63^\circ$$

$$i_s(t) = \underline{\underline{460.7 \cos(2000t + 52.63^\circ) \text{ mA}}}$$

Q6. Problem 9.56



$$50\mu F \longrightarrow \frac{1}{j\omega C} = \frac{1}{j377 \times 50 \times 10^{-6}} = -j53.05$$

$$60mH \longrightarrow j\omega L = j377 \times 60 \times 10^{-3} = j22.62$$

$$Z_{in} = 12 - j53.05 + j22.62 // 40 = \underline{\underline{21.692 - j35.91 \Omega}}$$