

EECS70A / CSE 70A Network Analysis I  
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Homework # 6 solution

**Chapter 8, Solution 17.**

$$i(0) = I_0 = 0, \quad v(0) = V_0 = 4 \times 15 = 60$$

$$\frac{di(0)}{dt} = -\frac{1}{L}(RI_0 + V_0) = -4(0 + 60) = -240$$

$$\omega_o = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{\frac{1}{4} \frac{1}{25}}} = 10$$

$$\alpha = \frac{R}{2L} = \frac{10}{2 \cdot \frac{1}{4}} = 20, \text{ which is } > \omega_o.$$

$$s = -\alpha \pm \sqrt{\alpha^2 - \omega_o^2} = -20 \pm \sqrt{300} = -20 \pm 10\sqrt{3} = -2.679, -37.32$$

$$i(t) = A_1 e^{-2.679t} + A_2 e^{-37.32t}$$

$$i(0) = 0 = A_1 + A_2, \quad \frac{di(0)}{dt} = -2.679A_1 - 37.32A_2 = -240$$

$$\text{This leads to } A_1 = -6.928 = -A_2$$

$$i(t) = 6.928 \left( e^{-37.32t} - e^{-2.679t} \right)$$

$$\text{Since, } v(t) = \frac{1}{C} \int_0^t i(t) dt + \text{const, and } v(0) = 60V, \text{ we get}$$

$$v(t) = \underline{\underline{(64.65e^{-2.679t} - 4.641e^{-37.32t}) V}}$$

**We note that  $v(0) = 60.009V$  and not  $60V$ . This is due to rounding errors since  $v(t)$  must go to zero as time goes to infinity. {In other words, the constant of integration must be zero.**

**Chapter 8, Solution 24**

When the switch is in position A, the inductor acts like a short circuit so

$$i(0^-) = 4$$

When the switch is in position B, we have a source-free parallel RCL circuit

$$\alpha = \frac{1}{2RC} = \frac{1}{2 \times 10 \times 10 \times 10^{-3}} = 5$$

$$\omega_o = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{\frac{1}{4} \times 10 \times 10^{-3}}} = 20$$

Since  $\alpha < \omega_o$ , we have an underdamped case.

$$s_{1,2} = -5 + \sqrt{25 - 400} = -5 + j19.365$$

$$i(t) = e^{-5t} (A_1 \cos 19.365t + A_2 \sin 19.365t)$$

$$i(0) = 4 = A_1$$

$$v = L \frac{di}{dt} \longrightarrow \frac{d i(0)}{dt} = \frac{v(0)}{L} = 0$$

$$\frac{di}{dt} = e^{-5t} (-5A_1 \cos 19.365t - 5A_2 \sin 19.365t - 19.365A_1 \sin 19.365t + 19.365A_2 \cos 19.365t)$$

$$0 = \frac{d i(0)}{dt} = -5A_1 + 19.365A_2 \longrightarrow A_2 = \frac{5A_1}{19.365} = 1.033$$

$$i(t) = \underline{e^{-5t} (4 \cos 19.365t + 1.033 \sin 19.365t)}$$

### Chapter 8, Solution 34.

Before  $t = 0$ , the capacitor acts like an open circuit while the inductor behaves like a short circuit.

$$i(0) = 0, v(0) = 20 \text{ V}$$

For  $t > 0$ , the LC circuit is disconnected from the voltage source.

This is a lossless, source-free, series RLC circuit.

$$\alpha = R/(2L) = 0, \omega_o = 1/\sqrt{LC} = 1/\sqrt{\frac{1}{16} + \frac{1}{4}} = 8, s = \pm j8$$

Since  $\alpha$  is less than  $\omega_0$ , we have an underdamped response. Therefore,

$$i(t) = A_1 \cos 8t + A_2 \sin 8t \quad \text{where } i(0) = 0 = A_1$$

$$di(0)/dt = (1/L)v_L(0) = -(1/L)v(0) = -4 \times 20 = -80$$

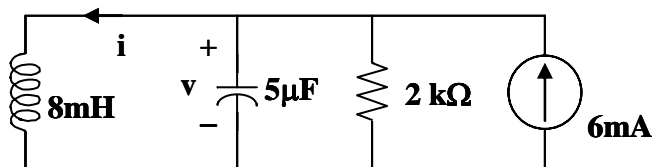
However,  $di/dt = 8A_2 \cos 8t$ , thus,  $di(0)/dt = -80 = 8A_2$  which leads to  $A_2 = -10$

Now we have  $i(t) = \underline{\underline{-10 \sin 8t \text{ A}}}$

### Chapter 8, Solution 46.

For  $t = 0^-$ ,  $u(t) = 0$ , so that  $v(0) = 0$  and  $i(0) = 0$ .

For  $t > 0$ , we have a parallel RLC circuit with a step input



$$\alpha = 1/(2RC) = (1)/(2 \times 2 \times 10^3 \times 5 \times 10^{-6}) = 50$$

$$\omega_0 = 1/\sqrt{LC} = 1/\sqrt{8 \times 10^{-3} \times 5 \times 10^{-6}} = 5,000$$

Since  $\alpha < \omega_0$ , we have an underdamped response.

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_0^2} \cong -50 \pm j5,000$$

Thus,  $i(t) = I_s + [(A \cos 5,000t + B \sin 5,000t)e^{-50t}]$ ,  $I_s = 6\text{mA}$

$$i(0) = 0 = 6 + A \quad \text{or } A = -6\text{mA}$$

$$v(0) = 0 = L di(0)/dt$$

$$di/dt = [5,000(-A \sin 5,000t + B \cos 5,000t)e^{-50t}] + [-50(A \cos 5,000t + B \sin 5,000t)e^{-50t}]$$

$$di(0)/dt = 0 = 5,000B - 50A \quad \text{or } B = 0.01(-6) = -0.06\text{mA}$$

Thus,  $i(t) = \underline{\underline{\{6 - [(6 \cos 5,000t + 0.06 \sin 5,000t)e^{-50t}]\} \text{ mA}}}$

**Chapter 8, Solution 57.**

(a) Let  $v$  = capacitor voltage and  $i$  = inductor current. At  $t = 0^-$ , the switch is closed and the circuit has reached steady-state.

$$v(0^-) = 16\text{V} \quad \text{and} \quad i(0^-) = 16/8 = 2\text{A}$$

At  $t = 0^+$ , the switch is open but,  $v(0^+) = 16$  and  $i(0^+) = 2$ .

We now have a source-free RLC circuit.

$$R = 8 + 12 = 20 \text{ ohms}, \quad L = 1\text{H}, \quad C = 4\text{mF}.$$

$$\alpha = R/(2L) = (20)/(2 \times 1) = 10$$

$$\omega_o = 1/\sqrt{LC} = 1/\sqrt{1 \times (1/36)} = 6$$

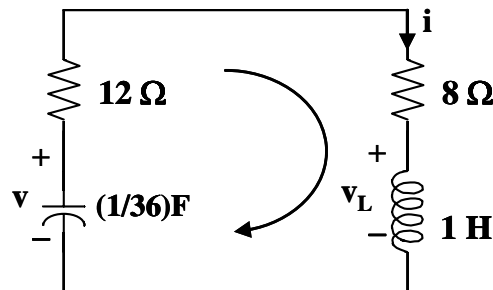
Since  $\alpha > \omega_o$ , we have an overdamped response.

$$s_{1,2} = -\alpha \pm \sqrt{\alpha^2 - \omega_o^2} = -18, -2$$

Thus, the characteristic equation is  $(s + 2)(s + 18) = 0$  or  $s^2 + 20s + 36 = 0$ .

$$(b) \quad i(t) = [Ae^{-2t} + Be^{-18t}] \quad \text{and} \quad i(0) = 2 = A + B \quad (1)$$

To get  $di(0)/dt$ , consider the circuit below at  $t = 0^+$ .



$$-v(0) + 20i(0) + v_L(0) = 0, \text{ which leads to,}$$

$$-16 + 20 \times 2 + v_L(0) = 0 \text{ or } v_L(0) = -24$$

$$L \frac{di(0)}{dt} = v_L(0) \text{ which gives } \frac{di(0)}{dt} = \frac{v_L(0)}{L} = \frac{-24}{1} = -24 \text{ A/s}$$

$$\text{Hence } -24 = -2A - 18B \text{ or } 12 = A + 9B \quad (2)$$

$$\text{From (1) and (2), } B = 1.25 \text{ and } A = 0.75$$

$$i(t) = [0.75e^{-2t} + 1.25e^{-18t}] = -i_x(t) \text{ or } i_x(t) = \underline{[-0.75e^{-2t} - 1.25e^{-18t}] \text{ A}}$$

$$v(t) = 8i(t) = \underline{[6e^{-2t} + 10e^{-18t}] \text{ A}}$$